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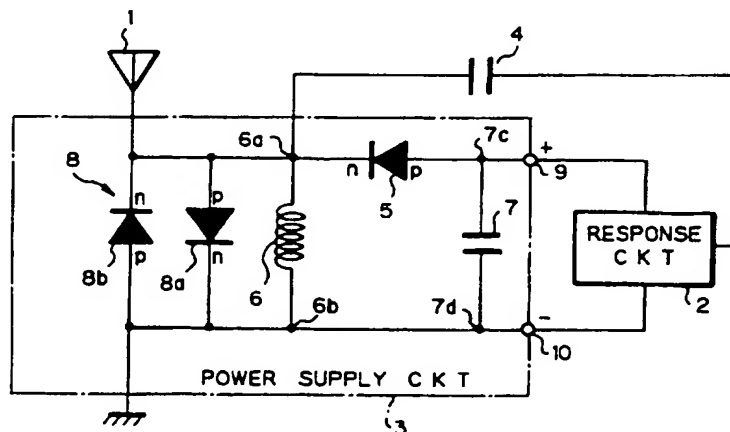
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(54) Power supply for a transponder

(57) A power supply 3 for a response circuit 2 of a transponder generates DC power from received electromagnetic energy and includes a capacitor 7 charged from a rectifier diode 5 which has a characteristic such that its reverse resistance against a reverse current directed from its n region to its p region is lower than its forward resistance against a forward current from the p region to the n region. The diode 5 is thus connected reversely compared to a conventional diode, its anode being connected to the positive plate of capacitor 7. The arrangement allows the transponder to remain functional even when the received electromagnetic energy is weak. The required characteristic for diode 5 can be implemented by the avalanche or tunnel effect. The diode 5 may be formed from a compound semiconductor or an organic semiconductor. The electromagnetic energy may be received by an antenna 1, or the transponder may be coupled to an interrogator by electromagnetic coupling, electromagnetic induction or microwave. A limiter 8 formed by ordinary diodes 8a, 8b may be provided to limit the voltage applied to the diode 5. A full wave rectifier arrangement may be provided using two capacitors and two diodes of the defined characteristic (Fig.5), or using a bridge rectifier and two receiving antennas (Fig.6), or a voltage multiplier may be provided (Figs.7,8).

Fig. 1



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Fig. 1

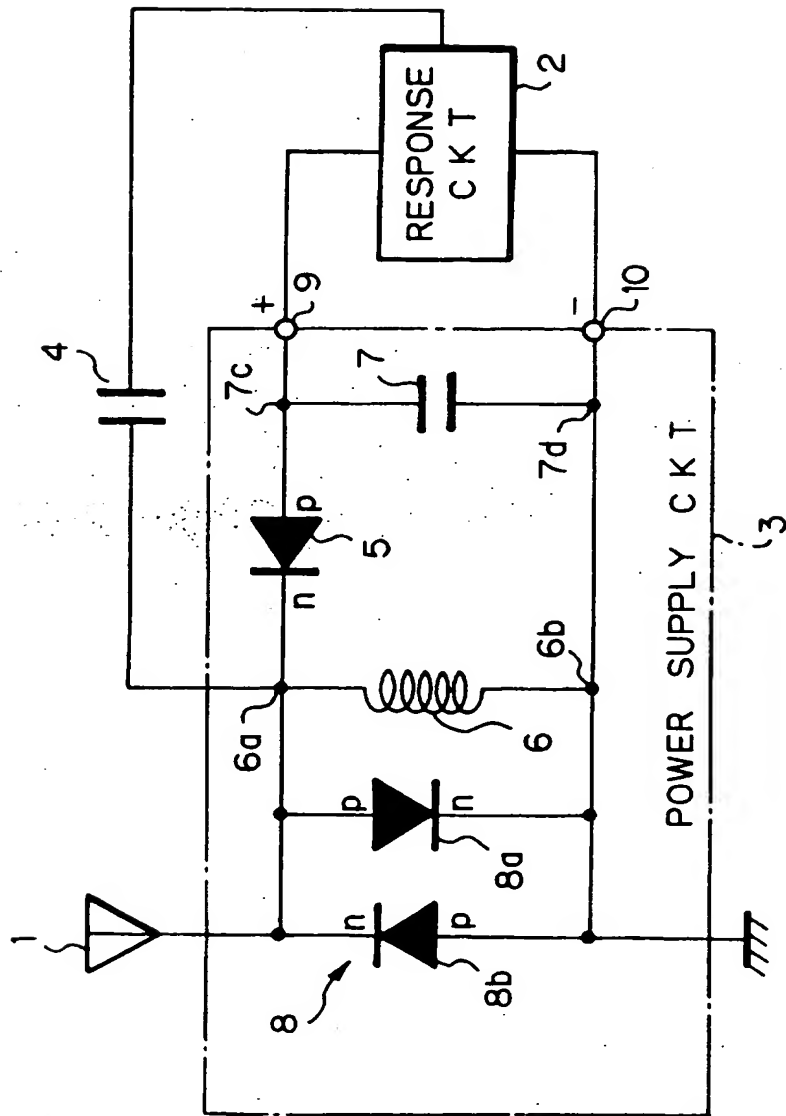
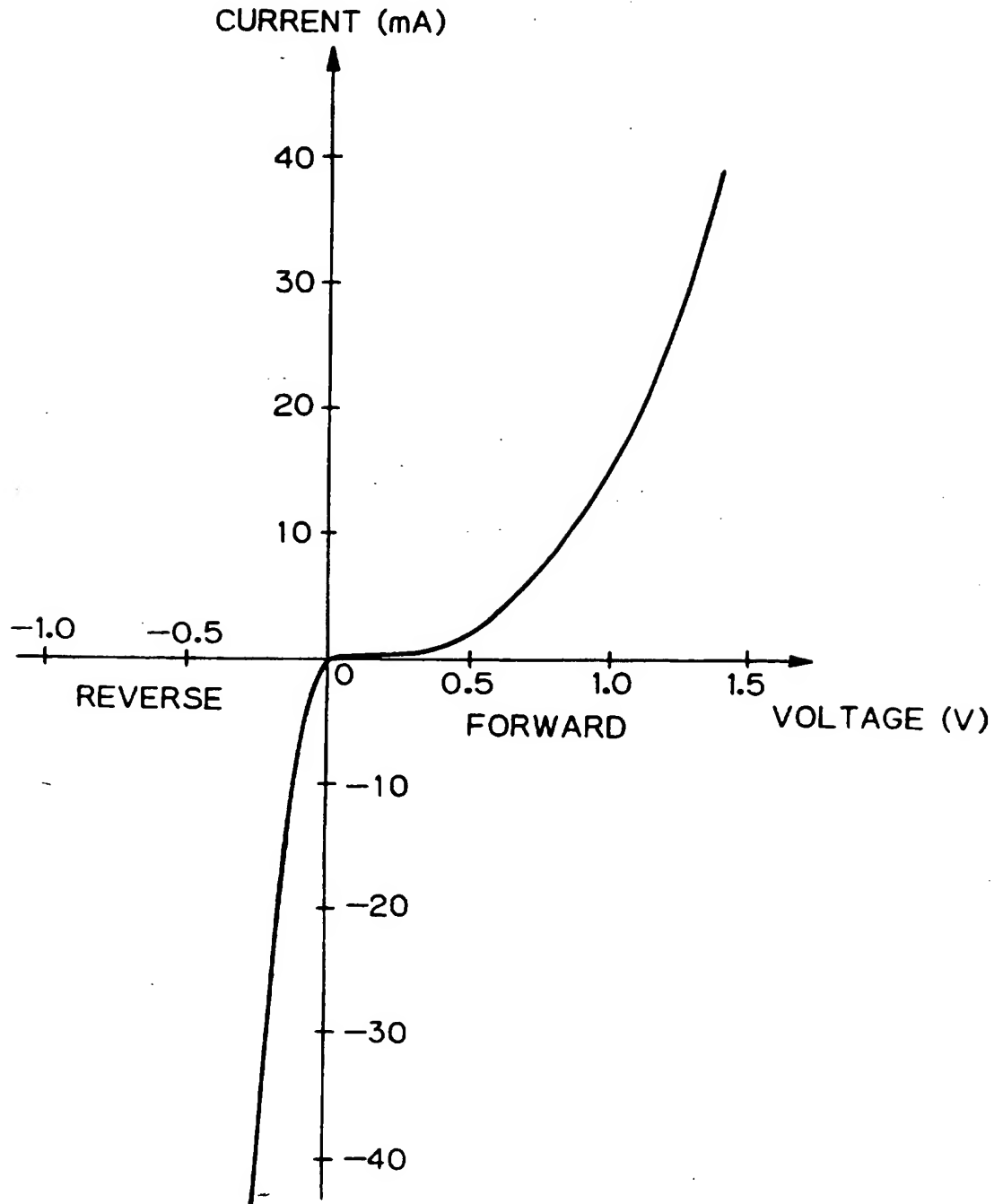


Fig. 2

CURRENT-TO-VOLTAGE CHARACTERISTIC
OF RECTIFIER DIODE

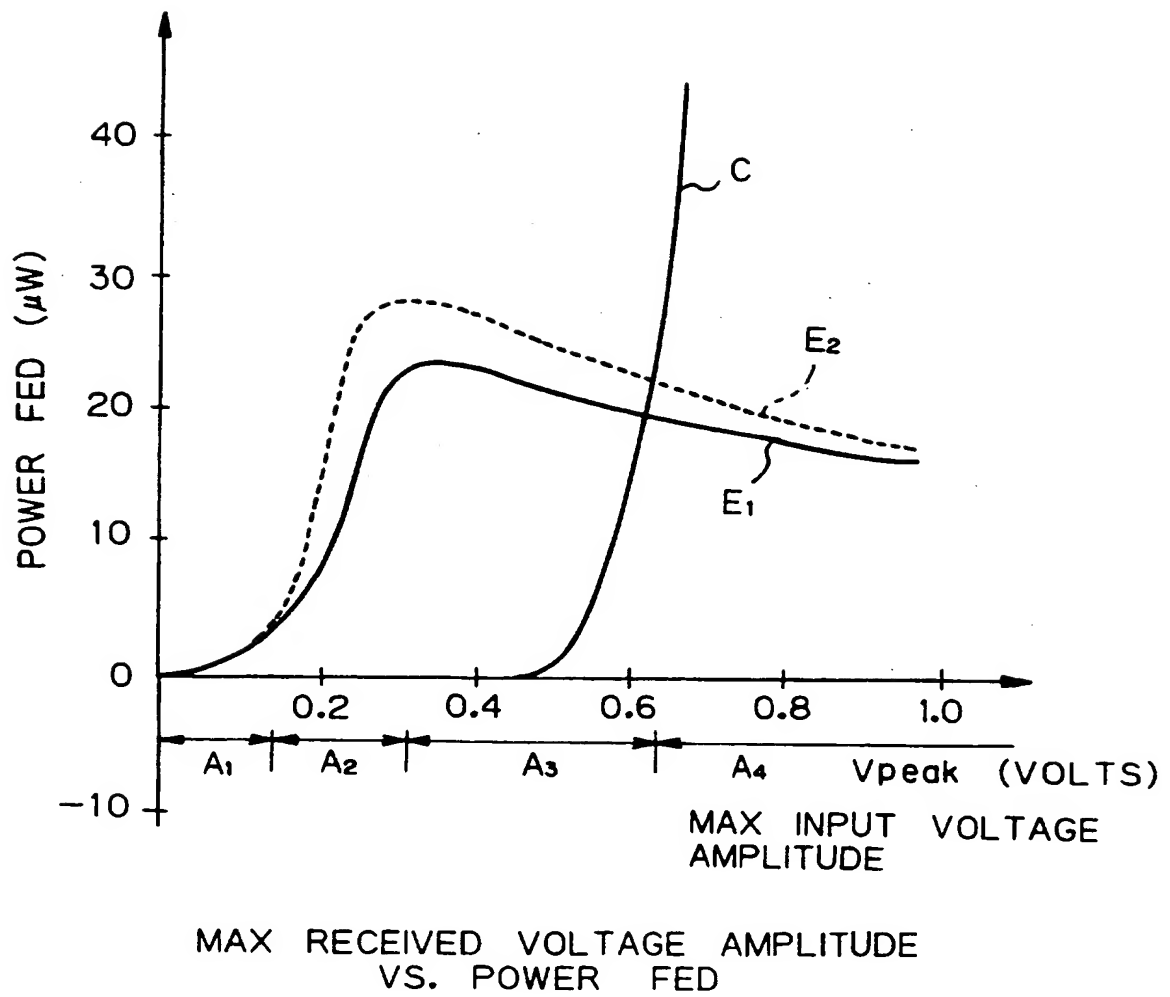
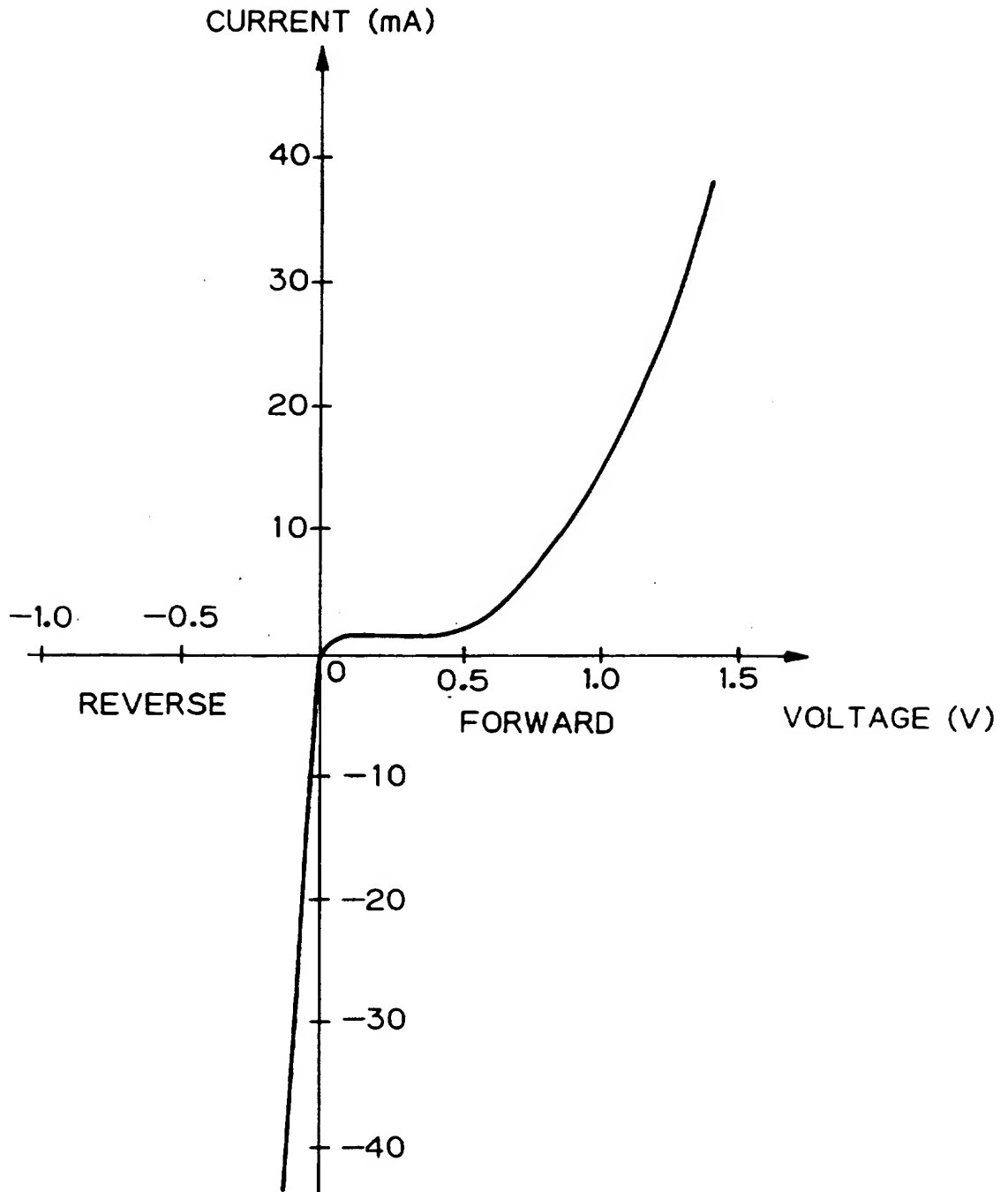
Fig. 3

Fig. 4

CURRENT-TO-VOLTAGE CHARACTERISTIC
OF RECTIFIER DIODE

Fig. 5

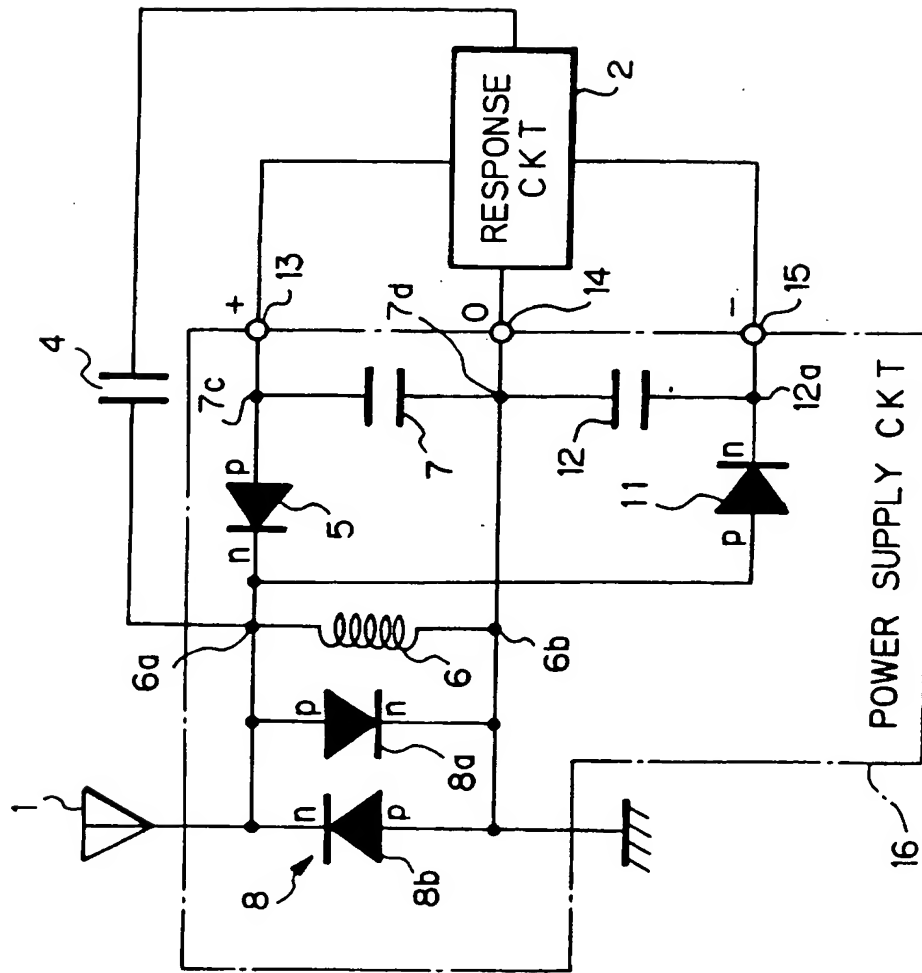


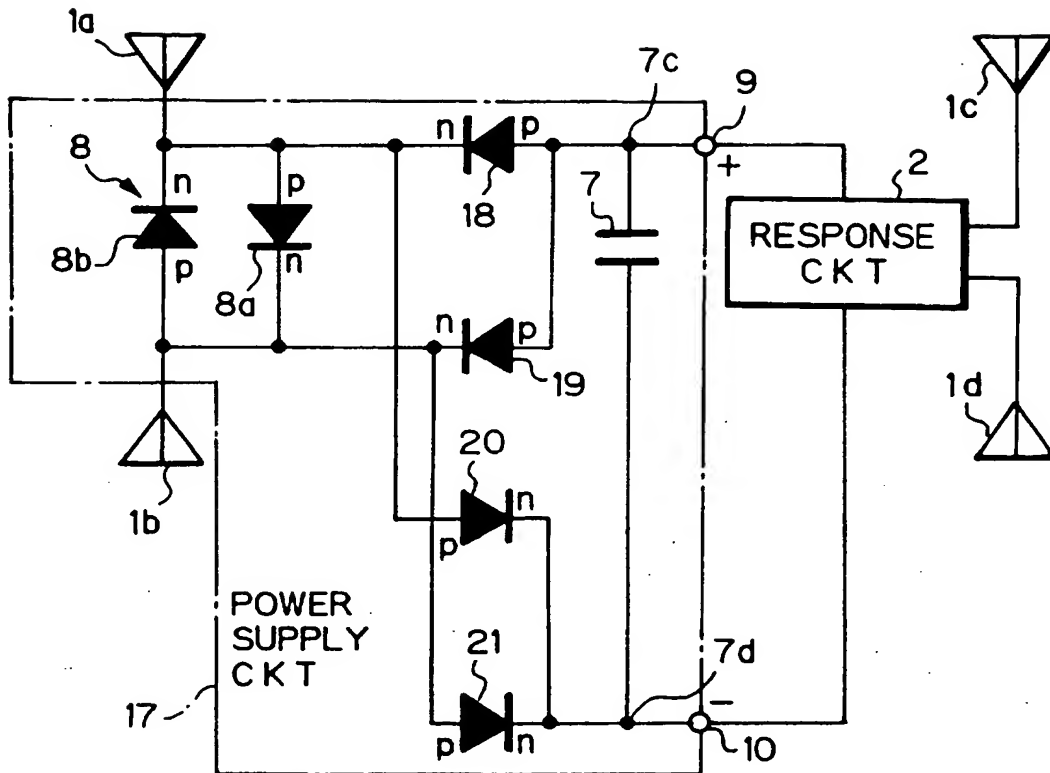
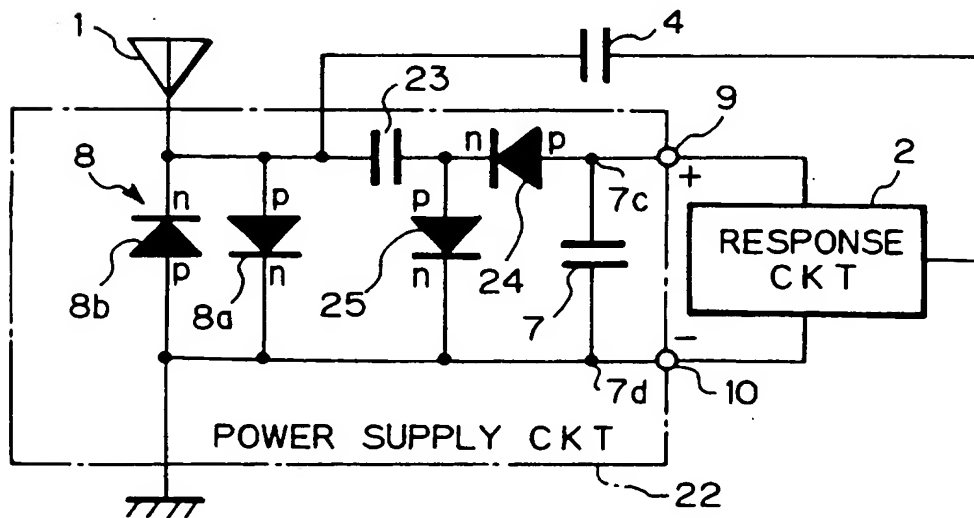
Fig. 6*Fig. 7*

Fig. 8

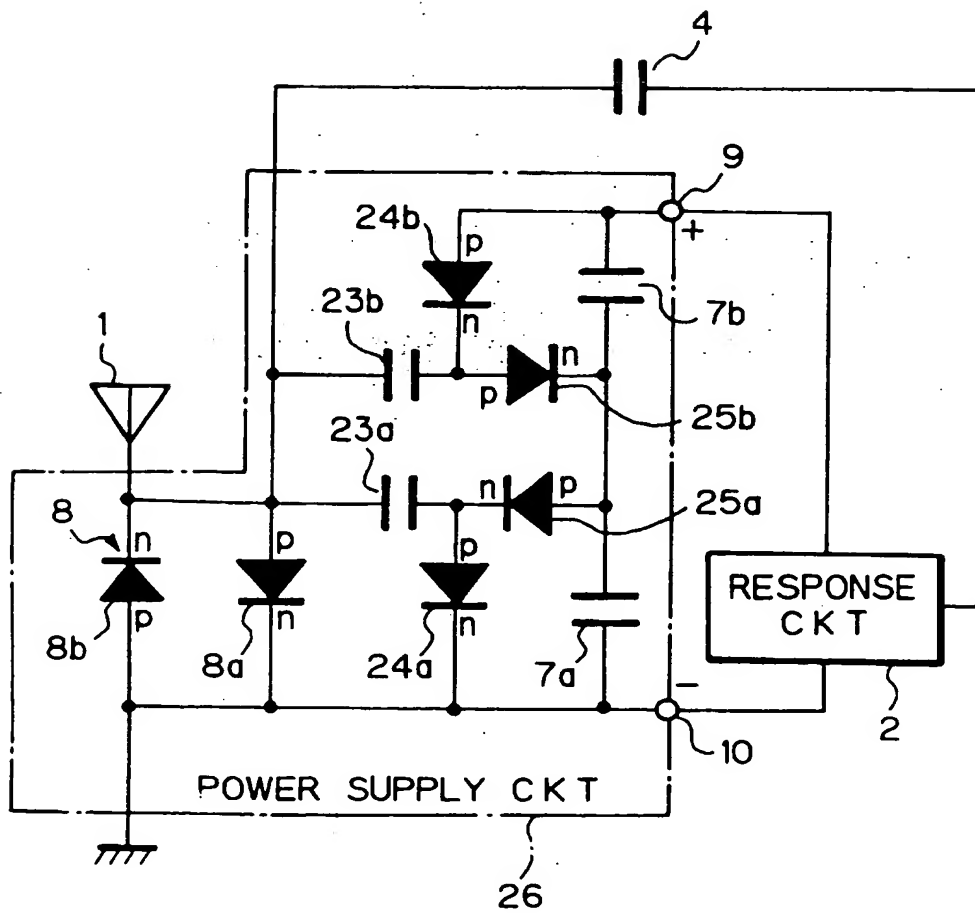


Fig. 9

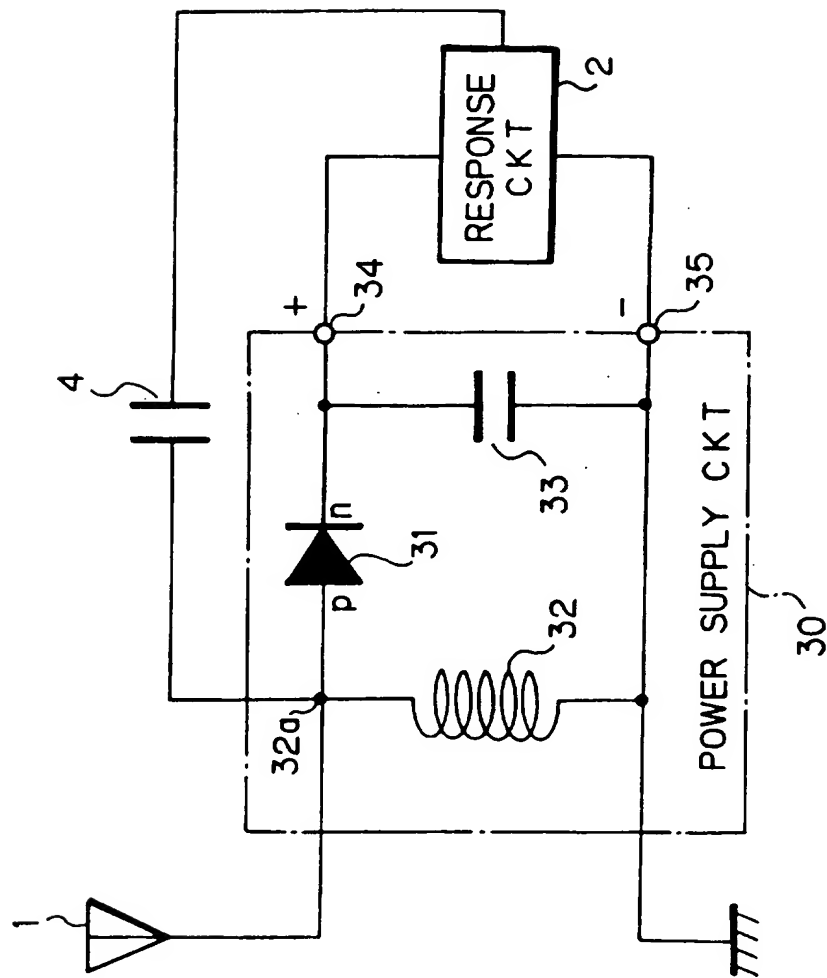
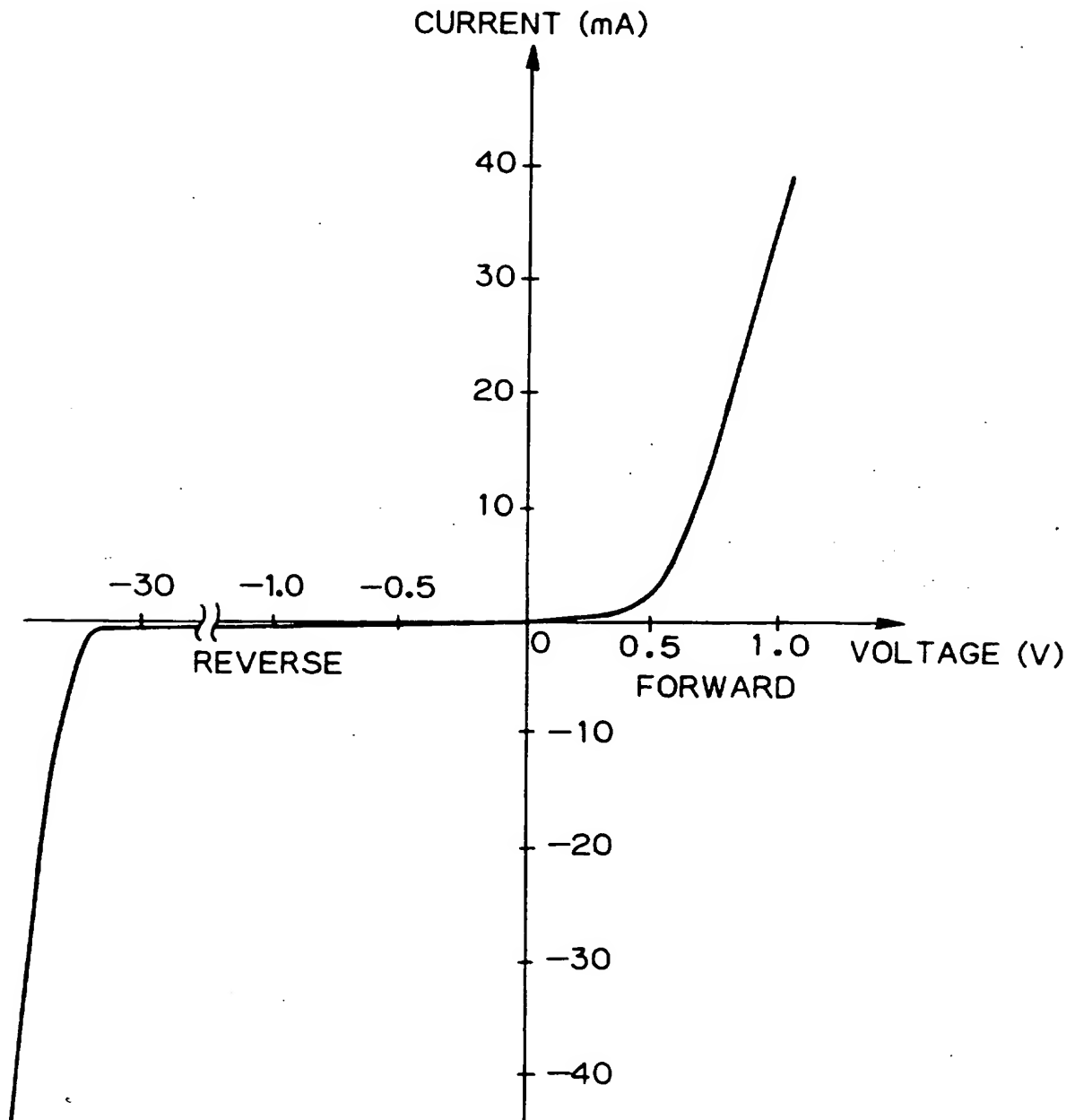


Fig. 10

CURRENT-TO-VOLTAGE CHARACTERISTIC
OF RECTIFIER DIODE

1 POWER SUPPLY CIRCUITRY FOR A TRANSPONDER AND
OPERABLE WITH INFINITESIMAL POWER

5
The present invention relates to power supply circuitry for a
transponder and capable of generating out of power received from an
interrogator DC power to be fed to a response circuit which outputs
10 information relating to, e.g., the kind and cost of an article in response to the
interrogator.

15 EP Publication 0 585 132 A1, for example, discloses a transponder
of the type receiving an electromagnetic wave from an interrogator with an
antenna, causing its response circuit to output information in response to
the received wave, and sending the information to the interrogator via the
antenna. Power supply for the response circuit is implemented by the
20 power of the received electromagnetic wave. For this purpose, power
supply circuitry has a flywheel coil for transforming the AC power of the
received wave to an AC electromotive force. The electromotive force is
rectified by a rectifier diode and then smoothed by a smoothing capacitor.
The resulting DC power is fed to the response circuit.

25
Specifically, on receiving the electromagnetic wave from the
interrogator via the antenna, the power supply circuitry transforms the
power of the received wave to the DC power and feeds the DC power to the
response circuit. The DC power enables or activates the response circuit.
30 On the receipt of an interrogation signal from the interrogator, the response

1 circuit demodulates and analyzes it and then outputs information
corresponding to the interrogation and stored therein. The information is
routed through a coupling capacitor to the antenna and sent to the
interrogator via the antenna.

5

To enable the power supply circuitry of the type described, it is
necessary to render the rectifier diode conductive. This cannot be done
unless an AC voltage having an amplitude capable of causing a forward
current to flow through the rectifier diode is input to the p type
10 semiconductor region of the diode. The minimum voltage amplitude
capable of causing the forward current to flow through the diode is
dependent on the kind of the diode. For example, the minimum voltage
amplitude is higher than about 0.6 V in the case of a silicon diode, or higher
than about 0.4 V in the case of a silicon Schottky barrier diode, or higher
15 than about 0.15 V in the case of a germanium diode. It has been customary
with power supply circuitry for a transponder to use a rectifier diode having
such a relatively high forward conduction voltage. This, however, brings
about a problem that to enable the circuitry for feeding power to the
response circuit, the maximum amplitude of the voltage of the received
20 signal input to the circuitry must be increased by intensifying the
electromagnetic wave to be radiated from the interrogator or by using an
antenna having a high voltage gain.

25 It is therefore an object of the present invention to provide power
supply circuitry for a transponder and capable of feeding DC power to its
response circuit even if an electromagnetic wave received from an
interrogator is weak or its antenna gain is low.

30 In accordance with the present invention, power supply circuitry

1 for a transponder is connected to a receiving section for receiving
electromagnetic energy and a response circuit connected to the receiving
section for transmitting a signal in response to received electromagnetic
energy. The power supply circuitry generates DC power to be fed to the
5 response circuit out of the received electromagnetic energy. The power
supply circuitry has a first rectifier diode having a junction formed by an n
type semiconductor region and a p type semiconductor region. The first
rectifier diode is adjusted such that a reverse resistance against a reverse
current directed from the n type semiconductor region toward the p type
10 semiconductor region is lower than a forward resistance against a forward
current directed from the p type semiconductor region toward the n type
semiconductor region. The first rectifier diode is connected between the
receiving section and the response circuit. A voltage output from the
receiving section received the electromagnetic energy is applied to the first
15 rectifier diode to thereby cause the reverse current to flow through the first
rectifier diode. The resulting DC voltage output from the first rectifier
diode causes power to be fed to the response circuit.

Also, in accordance with the present invention, a transponder has a
20 receiving section for receiving electromagnetic energy sent from an
interrogator. A response circuit is connected to the receiving section for
transmitting a signal to the interrogator in response to the electromagnetic
energy received. Power supply circuitry is connected to said receiving
section and response circuit for generating DC power to be applied to the
25 response circuit out of the electromagnetic energy received. The power
supply circuitry has a first rectifier diode having a junction formed by an n
type semiconductor region and a p type semiconductor region. The first
rectifier diode is adjusted such that a reverse resistance against a reverse
current directed from the n type semiconductor region toward the p type
30 semiconductor region is lower than a forward resistance against a forward

1 current directed from the p type semiconductor region toward the n type
semiconductor region. The first rectifier diode is connected between the
receiving section and the response circuit. A voltage output from the
receiving section received the electromagnetic energy is applied to the first
5 rectifier diode to thereby cause the reverse current to flow through the first
rectifier diode. The resulting DC voltage output from the first rectifier
diode causes power to be fed to the response circuit.

10 The objects and features of the present invention will become more
apparent from the consideration of the following detailed description taken
in conjunction with the accompanying drawings in which:

FIG. 1 is a schematic circuit diagram showing a transponder
including a preferred embodiment of the power supply circuitry in
15 accordance with the present invention;

FIG. 2 is a graph showing a specific current-to-voltage
characteristic of a rectifier diode included in the embodiment shown in FIG.
1;

FIG. 3 is a graph comparing the embodiment of FIG. 1 and a
20 conventional transponder with respect to a relation between the maximum
amplitude of a voltage and power actually fed;

FIG. 4 is a graph showing another specific current-to-voltage
characteristic of the rectifier diode included in the embodiment of FIG. 1;

FIG. 5 is a schematic circuit diagram showing a transponder
25 including an alternative embodiment of the present invention;

FIG. 6 is a schematic circuit diagram showing a transponder
including another alternative embodiment of the present invention;

FIG. 7 is a schematic circuit diagram showing a transponder
including still another alternative embodiment of the present invention;

30 FIG. 8 is a schematic circuit diagram showing a transponder

1 including a further alternative embodiment of the present invention;

FIG. 9 is a schematic circuit diagram showing a conventional transponder; and

5 FIG. 10 is a graph representative of a specific current-to-voltage characteristic particular to an ordinary diode.

Referring to FIG. 1 of the drawings, a transponder including power supply circuitry embodying the present invention is shown. As shown, the
10 transponder has a transmit/receive antenna 1 for receiving an electromagnetic wave sent from an interrogator, not shown, and for sending a signal to the interrogator. A response circuit 2 outputs, in response to the received electromagnetic wave or signal, a signal based on information stored therein. The power supply circuitry, generally 3, is also responsive
15 to the received signal for generating DC power to be fed to the response circuit 2. The signal output from the response circuit 2 is routed through a coupling capacitor 4 to the antenna 1 and radiated via the antenna 1.

The circuitry 3 includes a reverse diode or similar rectifier diode 5
20 having a junction formed by an n type and a p type semiconductor region. The rectifier diode 5 is adjusted such that a reverse resistance against a reverse current directed from the n type semiconductor region toward the p type semiconductor region is lower than a forward resistance against a forward current directed from the latter region toward the former region.
25 Assume a reverse and a forward resistance against a reverse and a forward current, respectively, having the same value at least smaller than the mean load current of the response circuit 2. Then, the reverse resistance of the diode 5 is lower than the forward resistance of the same. The reverse conduction mechanism of the diode 5 may be implemented by the
30 conventional avalanche effect or the tunnel effect. The n type region of the

1 diode 5 is connected to the antenna 1 while the p type region of the diode 5 is
connected to the response circuit 2 via a terminal 9.

5 Generally, a semiconductor diode has a forward voltage drop
characteristic dependent on the kind of a semiconductor used. For example,
a silicon diode has such a characteristic that the forward current sharply
increases when a forward voltage applied to the diode rises above about 0.6
V, as shown in FIG. 10 specifically. On the other hand, the semiconductor
diode can be provided with any desired reverse voltage drop characteristic if
10 its impurity content is adjusted on the production line. FIG. 2 shows a
specific current-to-voltage characteristic available with the rectifier diode 5
included in the power supply circuit 3. As shown, the reverse breakdown
voltage is close to 0 V. This kind of semiconductor diode, as distinguished
from an "ordinary" diode whose forward resistance is lower than the reverse
15 resistance, can be produced if the impurity content is increased in both the p
type region and the n type region so as to set up a field strength of the p - n
junction higher than that of the "ordinary" diode on the application of a
reverse current.

20 Referring again to FIG. 1, the circuitry 3 further includes a
flywheel coil 6 and a smoothing capacitor 7. The flywheel coil 6 allows a
pulsation derived from the rectification of the diode 5 to pass therethrough.
The smoothing capacitor 7 smoothes the rectified voltage and stores the
resulting power therein. The coil 6 is connected at one end 6a to the n type
25 region of the rectifier diode 5 and at the other end 6b to a reference potential,
e.g., ground. The capacitor 7 is connected at one end 7c to the p type region
of the rectifier diode 5 and at the other end to the above reference potential,
e.g., ground.

30 A limiter circuit 8 has two ordinary diodes 8a and 8b each having a

1 forward resistance lower than a reverse resistance, as stated earlier. The
limiter circuit 8 limits the voltage to be applied to the n type region of the
rectifier diode 5. Basically, the diodes or protection diodes 8a and 8b may
5 have an identical current-to-voltage characteristic, e.g., the characteristic
shown in FIG. 10; the forward current sharply increases when the forward
voltage rises above about 0.6 V. The protection diode 8a has its p type
semiconductor region and n type semiconductor region connected to the n
type region of the rectifier diode 5 and ground, respectively. The other
10 protection diode 8 has its n type semiconductor region and p type
semiconductor region connected to the n type region of the rectifier diode 5
and ground, respectively.

In operation, an electromagnetic wave radiated from the
interrogator is received by the antenna 1 and causes an AC electromotive
15 force to occur at one end 6a of the flywheel coil 6. The electromotive force
causes a reverse current and a forward current smaller than the reverse
current to flow alternately. As a result, a pulsation corresponding to a
difference between the reverse and forward currents is generated. The
pulsation is smoothed by the smoothing capacitor 7. Consequently, a
20 positive (+) and a negative (-) DC voltage appear on the terminal 9 and a
terminal 10, respectively. The response circuit 2 is enabled by the positive
and negative DC voltages. Specifically, the response circuit 2 demodulates
and analyzes the received signal and outputs information corresponding to
the received signal and stored therein. The information is routed through
25 the coupling capacitor 4 to the antenna 1 and radiated in the form of an
electromagnetic wave via the antenna 1.

FIG. 4 shows another specific current-to-voltage characteristic of
the rectifier diode 5. As shown, the diode 5 is again adjusted such that the
30 reverse resistance against the reverse current directed from the n type

1 region toward the p type region is lower than the forward current directed
from the latter region to the former region . While the characteristic
described with reference to FIG. 2 has a reverse breakdown voltage of about
0 V, the characteristic shown in FIG. 4 has a breakdown voltage at the
5 positive side (about 0.1 V in FIG. 4).

The illustrative embodiment will be compared with a conventional
transponder shown in FIG. 9. As shown, the conventional transponder has
power supply circuitry 30 in addition to the transmit/receive antenna 1,
10 response circuit 2, and coupling capacitor 4. The antenna 1 receives an
electromagnetic wave from an interrogator, not shown, and transmits an
electromagnetic wave to the interrogator. The response circuit 2 outputs,
in response to the received signal, a signal based on information stored
therein. The circuitry 30 feeds DC power to the response circuit 2. The
15 signal output from the response circuit 2 is fed to the antenna 1 via the
coupling capacitor 4. The circuitry 30 is made up of a rectifier diode 31, a
flywheel coil 32 and a smoothing capacitor 33 connected together, as
illustrated. The rectifier diode 31 has its p type semiconductor region
connected to the antenna 1. The flywheel coil 32 allows a pulsation derived
20 from the rectification of the diode 31 to pass therethrough. The smoothing
capacitor 33 smoothes the rectified voltage. The diode 31 is implemented
by the previously mentioned "ordinary" diode having the current-to-voltage
characteristic shown in FIG. 10.

25 The operation of the conventional transponder is as follows. An
electromagnetic wave radiated from interrogator is received by the antenna
1 and causes an AC electromotive force to occur at one end 32a of the
flywheel coil 32. The AC electromotive force is rectified by the rectifier
diode 31. A rectified voltage output from the rectifier 31 is smoothed by the
30 smoothing capacitor 33. As a result, a positive and a negative DC voltage

1 respectively appear on terminals 34 and 35 and enable the response circuit
2 The response circuit 2 demodulates and analyzes the received signal fed
from the antenna 1 and outputs corresponding information stored therein.
The information is delivered to the antenna 1 via the coupling capacitor 4.

5

FIG. 3 is a graph comparing the illustrative embodiment and the
conventional power supply circuit 30 (FIG. 9) with respect to a relation
between the maximum amplitude V_{peak} (V) of the voltage input to the
antenna 1 and the power (μW) fed to the response circuit 2. In FIG. 3,
10 curves E_1 and E_2 are each representative of the characteristic of the
embodiment while a curve C is representative of the characteristic of the
conventional circuitry 30. As the curve C indicates, in ranges A_1 and A_2 in
which the maximum amplitude V_{peak} of the voltage input to the antenna 1 is
low, the conventional circuitry 30 causes hardly any forward current to flow
15 through the rectifier diode 31, and therefore causes hardly any power to be
input to the response circuit 2. By contrast, as the curve E_1 indicates, the
power supply circuitry 3 (FIG. 2) whose rectifier diode 5 has the
characteristic shown in FIG. 2 allows a reverse current to flow through the
diode 5 even in the range A_2 in which the maximum amplitude V_{peak} is low.
20 It follows that the embodiment is capable of feeding sufficiently great DC
power to the response circuit 2.

As the curve E_1 further indicates, when the maximum amplitude of
the voltage input to the n type region of the rectifier diode 5 exceeds 0.3 V
25 (regions A_3 and A_4), power fed to the response circuit 2 decreases. This is
because when the maximum amplitude V_{peak} of the voltage input to the
antenna 1 rises above 0.3 V, the forward current flowing through the
rectifier diode 5 begins to increase. In the illustrative embodiment, the
limiter circuit 8 limits the amplitude of the voltage input to the n type region
30 of the diode 5. This successfully reduces the fall of the power ascribable to

1 the forward current flowing through the diode 5, even if the maximum
amplitude of the voltage output from the antenna 1 exceeds 0.3 V (regions
A₃ and A₄).

5 The curve E₂ shown in FIG. 3 is representative of another specific
characteristic of the rectifier diode 5 and derived from the voltage-to-current
characteristic shown in FIG. 4. It will be seen that with the characteristic
of FIG. 4 it is possible to feed greater power at large than with the
characteristic of FIG. 2.

10 The embodiment receives electromagnetic waves with its antenna 1
located in a radiation electromagnetic field. If desired, the antenna 1 may
be replaced with an arrangement which receives electromagnetic energy by
being coupled to the interrogator by electromagnetic coupling,
15 electromagnetic induction or microwave. The rectifier diode 5 may be
formed of a semiconductor other than silicon and having a great difference
between a forward resistance and a reverse resistance, e.g., a compound
semiconductor or an organic semiconductor.

20 Referring to FIG. 5, a transponder including an alternative
embodiment of the present invention will be described. In FIG. 1, the same
constituent parts as the constituent parts shown in FIG. 1 are designated by
like reference numerals, and a detailed description thereof will not be made
in order to avoid redundancy. As shown, power supply circuitry 16
25 includes a rectifier diode 11 in addition to the rectifier diode 5. The diode 5
has substantially the same characteristic as the diode 11, i.e., the specific
current-to-voltage characteristic shown in FIG. 2 or 4. The power supply
circuitry 16 further includes a capacitor 12 in addition to the capacitor 7.
The capacitors 7 and 12 are substantially identical in characteristic with
30 each other.

1 The rectifier diode 5 has its p type semiconductor region connected
to the response circuit 2 via a terminal 13 and has its n type semiconductor
region connected to the antenna 1. The other rectifier diode 11 has a p type
semiconductor region connected to the n type region of the diode 5 and has
5 an n type semiconductor region connected to the response circuit 12 via one
end 12a of the capacitor 12 and a terminal 15. The other end 7d of the
capacitor 12 is connected to ground and connected to the response circuit 2
via a terminal 14. The rectifier diodes 5 and 11 and capacitors 7 and 12
play the role of a full-wave double voltage rectifying circuit in combination.

10 In operation, an electromagnetic wave sent from an interrogator,
not shown, is received by the antenna 1 and causes an AC electromotive
force to occur at one end 6a of the flywheel coil 6. The AC electromotive
force causes a reverse current and a forward current smaller than the
15 reverse current to flow through one rectifier diode 11 alternately. As a
result, a pulsation corresponding to a difference between the reverse and
forward currents is generated. The pulsation is smoothed by one
smoothing capacitor 12 and turns out a negative DC voltage on the terminal
15. In response to the DC voltage, the response circuit 2 demodulates and
20 analyzes the received signal and outputs corresponding information stored
therein. The information is fed to the antenna 1 via the coupling capacitor
4 and radiated via the antenna 1. With this embodiment, it is possible to
feed sufficiently great DC power to the response circuit 2 even if the
maximum amplitude of the voltage received by the antenna 1 is small. As
25 for the rest of the construction, this embodiment may be identical with the
embodiment shown in FIG. 1.

FIG. 6 shows a transponder including another alternative
embodiment of the present invention. In FIG. 6, the same constituent

1 parts as the constituent parts shown in FIG. 1 are designated by like
reference numerals, and a detailed description thereof will not be made in
order to avoid redundancy. As shown, power supply circuitry 17 includes
5 four rectifier diodes 18, 19, 20 and 21 having substantially the same
characteristic, i.e., adjusted such that the reverse resistance is lower than
the forward resistance. The diodes 18-21 constitute a bridge rectifier
circuit, and each has the current-to-voltage characteristic shown in FIG. 2 or
4.

10 The transponder shown in FIG. 6 has two antennas 1a and 1b for
receipt and two antennas 1c and 1d for transmission. The receive antenna
1a is connected to the *n* type semiconductor region of the rectifier diode 18
and the *p* type semiconductor region of the rectifier diode 20. The other
receive antenna 1b is connected to the *n* type semiconductor region of the
15 rectifier diode 19 and the *p* type semiconductor region of the rectifier diode
21. The transmit antennas 1c and 1d are both connected to the output of
the response circuit 2.

In operation, an electromagnetic wave sent from an interrogator,
20 not shown, is received by the receive antennas 1a and 1b and causes an AC
electromotive force to occur. The AC electromotive force causes a reverse
current and a forward current smaller than the reverse current to flow
through the rectifier diodes 18-21. As a result, a pulsation corresponding
to a difference between the reverse and forward currents is generated. The
25 pulsation is smoothed by the smoothing capacitor 7 and causes a positive DC
voltage and a negative DC voltage to appear on the terminals 9 and 10,
respectively. In response to the DC voltages, the response circuit 2
demodulates and analyzes the received signal and outputs corresponding
information stored therein. The information is sent to the interrogator via
30 the transmit antennas 1c and 1d. This embodiment is also capable of

1 feeding sufficiently great DC power to the response circuit 2 even when the
maximum amplitude of the voltage input from the receive antennas 1a and
1b is small. As for the rest of the construction, this embodiment may also
be identical with the embodiment shown in FIG. 1.

5
FIG. 7 shows a transponder including still another alternative
embodiment of the present invention. In FIG. 7, the same constituent
parts as the constituent parts shown in FIG. 1 are designated by like
reference numerals, and a detailed description thereof will not be made in
10 order to avoid redundancy. As shown, power supply circuitry 22 has an
additional capacitor 23 and two rectifier diodes 24 and 25 having
substantially the same characteristic, i.e., adjusted such that the reverse
resistance is lower than the forward resistance. The rectifier diodes 24 and
25 have the current-to-voltage characteristic shown in FIG. 2 or 4 and
15 constitute a double voltage rectifying circuit.

In operation, an electromagnetic wave sent from an interrogator,
not shown, is received by the antenna 1 and causes an AC electromotive
force to occur. The AC electromotive force causes a reverse current and a
20 forward current smaller than the reverse current to flow through the
rectifier diodes 24 and 25. As a result, a pulsation corresponding to a
difference between the reverse and forward currents is generated. The
pulsation is smoothed by the smoothing capacitor 7 and causes a positive
and a negative DC current to appear on the terminals 9 and 10, respectively.
25 In response to the DC voltages, the response circuit 2 demodulates and
analyzes the received signal and outputs corresponding information stored
therein. The information is routed through the coupling capacitor 4 to the
antenna 1 and sent to the interrogator via the antenna 1. In this
configuration, even when the maximum amplitude of the voltage input from
30 the antenna 1 is small, sufficiently large DC power can be fed to the

1 response circuit 2, as in the previous embodiments. As for the rest of the construction, this embodiment may also be identical with the embodiment shown in FIG. 1.

5 FIG. 8 shows a transponder including a further alternative embodiment of the present invention. In FIG. 8, the same constituent parts as the constituent parts shown in FIG. 1 are designated by like reference numerals, and a detailed description thereof will not be made in order to avoid redundancy. As shown, power supply circuitry 26 has a pair
10 of additional capacitors 23a and 23b, two rectifier diodes 24a and 25a, and other two rectifier diodes 24b and 25b. The diodes 24a and 25a and the diodes 24b and 25b have substantially the same characteristic as each other and have the current-to-voltage characteristic shown in FIG. 2 or 4. In the configuration shown in FIG. 8, the diodes 24a and 25a and diodes 24b and
15 25b constitute a quadruple voltage rectifying circuit in combination with the capacitors 23a and 23b and capacitors 7a and 7b.

In operation, an electromagnetic wave sent from an interrogator, not shown, is received by the antenna 1 and causes an AC electromotive
20 force to occur. The AC electromotive force causes a reverse current and a forward current smaller than the reverse current to flow through the rectifier diodes 24a and 25a and rectifier diodes 24b and 25b. As a result, a pulsation corresponding to a difference between the reverse and forward currents is generated. The pulsation is smoothed by the smoothing
25 capacitor 7a and 7b and causes a positive and a negative DC current to appear on the terminals 9 and 10, respectively. In response to the DC voltages, the response circuit 2 demodulates and analyzes the received signal and outputs corresponding information stored therein. The information is routed through the coupling capacitor 4 to the antenna 1 and
30 sent to the interrogator via the antenna 1. In this configuration, even

1 when the maximum amplitude of the voltage input from the antenna 1 is
small, sufficiently large DC power can also be fed to the response circuit 2.
As for the rest of the construction, this embodiment may also be identical
with the embodiment shown in FIG. 1.

5 In summary, it will be seen that the present invention provides
power supply circuitry for a transponder and capable of feeding sufficient
DC power to a response circuit even when an electromagnetic wave received
from an interrogator is weak or when an antenna gain is low.

10 While the present invention has been described with reference to
the particular illustrative embodiments, it is not to be restricted by the
embodiments. It is to be appreciated that those skilled in the art can
change or modify the embodiments without departing from the scope and
15 spirit of the present invention.

CLAIMS

1 1. Power supply circuitry for a transponder and connected to
receiving means (1) for receiving electromagnetic energy and a response
circuit (2) connected to said receiving means (1) for transmitting a signal in
response to received electromagnetic energy, said power supply circuitry
5 generating DC power to be fed to said response circuit (2) out of the received
electromagnetic energy, CHARACTERIZED BY comprising:

 a first rectifier diode (5) having a junction formed by an *n* type
semiconductor region and a *p* type semiconductor region, and adjusted such
that a reverse resistance against a reverse current directed from said *n* type
10 semiconductor region toward said *p* type semiconductor region is lower than
a forward resistance against a forward current directed from said *p* type
semiconductor region toward said *n* type semiconductor region;

 said first rectifier diode (5) being connected between said receiving
means (1) and said response circuit (2);

15 whereby a voltage output from said receiving means received the
electromagnetic energy is applied to said first rectifier diode (5) to thereby
cause the reverse current to flow through said first rectifier diode (5), and a
resulting DC voltage output from said first rectifier diode (5) causes power
to be fed to said response circuit (2).

1 2. Power supply circuitry in accordance with claim 1,
CHARACTERIZED IN THAT said *n* type semiconductor region and said *p*
type semiconductor region of said first rectifier diode (5) are connected to
said receiving means (1) and said response circuit (2), respectively.

1 3. Power supply circuitry in accordance with claim 1,
CHARACTERIZED IN THAT said first rectifier diode (5) is adjusted such
that assuming a forward resistance and a reverse resistance against
currents having a same value within a range of a current to be fed to said

5 response circuit (2), said reverse resistance is lower than said forward resistance.

1 4. Power supply circuitry in accordance with claim 1, CHARACTERIZED IN THAT said first rectifier diode (5) comprises a silicon diode, and the reverse voltage of said first rectifier diode (5) is lower than 0.3 V inclusive.

1 5. Power supply circuitry in accordance with claim 1, CHARACTERIZED BY further comprising:

5 a coil (6) connected at one end to said *n* type semiconductor region of said first rectifier diode (5) and at the other end to a reference potential; and

a first capacitor (7) connected at one end (7c) to said *p* type semiconductor region of said first rectifier diode (5) and at the other end (7d) to said reference potential.

1 6. Power supply circuitry in accordance with claim 5, CHARACTERIZED BY further comprising:

a second rectifier diode (11) having substantially a same characteristic as said first rectifier diode (5); and

5 a second capacitor (12) having substantially a same characteristic as said first capacitor (7);

10 said second rectifier diode (11) comprising a junction formed by an *n* type semiconductor region and a *p* type semiconductor region, said *p* type semiconductor region and said *n* type semiconductor region being respectively connected to said *n* type semiconductor region of said first rectifier diode (5) and one end (7d) of said second capacitor (12);

the other end (12a) of said second capacitor (12) being connected to said reference potential.

1 7. Power supply circuitry in accordance with claim 1,
CHARACTERIZED BY further comprising a limiter circuit (8) for limiting a
voltage to be applied to said *n* type semiconductor region of said first
rectifier diode (5).

1 8. Power supply circuitry in accordance with claim 7,
CHARACTERIZED IN THAT said limiter circuit comprises a first protection
diode (8a) and a second protection diode (8b), said first protection diode (8a)
has a *p* type semiconductor region and an *n* type semiconductor region
5 thereof respectively connected to said *n* type semiconductor region of said
first rectifier diode (5) and said reference potential, and said second
protection diode (8b) has an *n* type semiconductor region and a *p* type
semiconductor region respectively connected to said *n* type semiconductor
region of said first rectifier diode (5) and said reference potential.

1 9. Power supply circuitry in accordance with claim 8,
CHARACTERIZED IN THAT said first protection diode (8a) and said
second protection diode (8b) each has an ordinary current-to-voltage
characteristic.

1 10. Power supply circuitry in accordance with claim 1,
CHARACTERIZED BY further comprising:

a capacitor (23) connected between said receiving means (1) and
said first rectifier diode (5); and

5 a second rectifier diode (25) connected between said capacitor (23)
and said reference potential and having substantially a same characteristic
as said first rectifier diode (5).

1 11. Power supply circuitry in accordance with claim 1,

CHARACTERIZED IN THAT said first rectifier diode (24b) is connected to a positive input (9) of said response circuit (2), said power supply circuit (26), and said power supply circuitry further comprises:

5 a first capacitor (23b) connected between said receiving means (1) and said first rectifier diode (24b);

 a series connection of a second capacitor (23a) and a second rectifier diode (24a) and connected between said receiving means (1) and a negative input (10) of said response circuit (2); and

10 a series connection of a third rectifier diode (25a) and a fourth rectifier diode (25b) and connected between a junction between said second capacitor (23a) and said second rectifier diode (24a) and a junction between said first capacitor (23b) and said first rectifier diode (24b), said third rectifier diode (25a) and said fourth rectifier diode (25b) having
15 substantially a same characteristic as said first rectifier diode.

1 12. Power supply circuitry in accordance with claim 1, CHARACTERIZED IN THAT said first rectifier diode (18) is connected to a positive input (9) of said response circuit (2), said power supply circuit (17), and said power supply circuitry further comprises:

5 a second rectifier diode (20) connected between said receiving means (1a) and a negative input (10) of said response circuit (2), and having substantially a same characteristic as said first rectifier diode (18), and connected in opposite direction to said first rectifier diode (18);

 a third rectifier diode (19) connected between additional receiving
10 means (1b) for receiving the electromagnetic energy and said positive input (9) of said response circuit (2), and having substantially a same characteristic as said first rectifier diode (18), and connected in a same direction as said first rectifier diode (18); and

 a fourth rectifier diode (21) connected between said additional
15 receiving means (1b) and said negative input (10) of said response circuit (2),

and having substantially a same characteristic as said third rectifier diode (19), and connected in opposite direction to said third rectifier diode (19).

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13. Power supply circuitry in accordance with claim 1,
CHARACTERIZED IN THAT said receiving means comprises an antenna
for receiving the electromagnetic energy in a form of an electromagnetic
wave.

BAD ORIGINAL



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Claims searched: 1 to 13

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Date of search: 11 October 1996

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Databases searched:

UK Patent Office collections, including GB, EP, WO & US patent specifications, in:

UK CI (Ed.O): H2F F9A; H4L LAPS.

Int CI (Ed.6): G01S 13/02, 13/76; G06K 19/07, 19/077; G07C 9/00; H02J 17/00;
H02M 7/06, 7/10; H04B 1/16.

Other: ONLINE : JAPIO, WPI.

Documents considered to be relevant:

Category	Identity of document and relevant passage	Relevant to claims
Y	US5305008 (INTEGRATED SILICON) - Fig.2A	1,2,13 at least
Y	US4912471 (MITRON) - Fig.4	1,2,7,13 at least
A	US4087791 (MINNESOTA MINING) - see diodes 42,44 in Fig.2	7,8
Y	A G Milnes, "Semiconductor Devices & Integrated Electronics", published 1980, Van Nostrand Reinhold, New York, see pages 178-179	1,2,7,13 at least

X Document indicating lack of novelty or inventive step
Y Document indicating lack of inventive step if combined with one or more other documents of same category.
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A Document indicating technological background and/or state of the art.
P Document published on or after the declared priority date but before the filing date of this invention.
E Patent document published on or after, but with priority date earlier than, the filing date of this application.

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